DIGITAL ARITHMETIC Miloš D. Ercegovac and Tomás Lang Morgan Kaufmann Publishers, an imprint of Elsevier Science, ©2004 – Updated: September 23, 2003 –

Chapter 6: Solutions to Exercises

- With contributions by Elisardo Antelo and Fabrizio Lamberti -

Exercise 6.1

a) Radix-2, $s_j \in \{-1, 0, 1\}$, conventional (nonredundant) residual

We have $x = 144 \times 2^{-8} = 0.10010000$ and $\rho = 1$. We choose $s_0 = 0$. Therefore the initialization is $w[0] = x - s_0 = 0.10010000$.

We use the result-digit selection function for redundant residual but we consider only 2 integer bits since the range of the residual estimate is smaller than in the redundant case.

2w[0] =	001.00100000	$\widehat{y} = 1$	$s_1 = 1$
$F_1[0] =$	11.10000000	$F_{-1}[0] = 11.10000000$	
w[1] =	00.10100000		
2w[1] =	001.01000000	$\widehat{y} = 1$	$s_2 = 1$
$F_1[1] =$	10.11000000	$F_{-1}[1] = 00.11000000$	
w[2] =	00.00000000		
2w[2] =	000.00000000	$\widehat{y} = 0$	$s_3 = 1$
$F_1[2] =$	10.01100000	$F_{-1}[2] = 01.01100000$	
w[3] =	10.01100000		
2w[3] =	100.11000000	$\widehat{y} = -4$	$s_4 = -1$
	$\begin{array}{c} 100.11000000\\ 01.10110000\end{array}$	$\hat{y} = -4$ $F_1[3] = 10.00110000$	$s_4 = -1$
$\begin{array}{c} 2w[3] = \\ F_{-1}[3] = \\ \hline w[4] = \end{array}$		0	$s_4 = -1$
	01.10110000	0	$s_4 = -1$
$\frac{F_{-1}[3] =}{w[4] =}$ $2w[4] =$	01.10110000	0	$s_4 = -1$ $s_5 = -1$
$ \begin{array}{c} F_{-1}[3] = \\ w[4] = \\ 2w[4] = \\ F_{-1}[4] = \\ \end{array} $	01.10110000 10.01110000	$F_1[3] = 10.00110000$	-
$\frac{F_{-1}[3] =}{w[4] =}$ $2w[4] =$	01.10110000 10.01110000 100.11100000	$\widehat{F}_1[3] = 10.00110000$ $\widehat{y} = -4$	-
$ \begin{array}{c} F_{-1}[3] = \\ w[4] = \\ 2w[4] = \\ F_{-1}[4] = \\ \end{array} $	01.10110000 10.01110000 100.11100000 01.10011000	$\widehat{F}_1[3] = 10.00110000$ $\widehat{y} = -4$	-
$ \frac{F_{-1}[3] =}{w[4] =} $ $ \frac{2w[4] =}{F_{-1}[4] =} $	01.10110000 10.01110000 100.11100000 01.10011000	$\widehat{F}_1[3] = 10.00110000$ $\widehat{y} = -4$	-
$\frac{F_{-1}[3] =}{w[4] =}$ $\frac{2w[4] =}{F_{-1}[4] =}$ $w[5] =$	$\begin{array}{c} 01.10110000\\ 10.01110000\\ 100.11100000\\ 01.10011000\\ 10.01111000\end{array}$	$\widehat{F}_{1}[3] = 10.00110000$ $\widehat{y} = -4$ $F_{1}[4] = 10.01011000$	$s_5 = -1$

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2w[6] =	100.11111000	$\widehat{y} = -4$	$s_7 = -1$
$F_{-1}[6] =$	01.10000110	$F_1[6] = 10.01110110$	
w[7] =	10.01111110		
2w[7] =	100.11111100	$\widehat{y} = -4$	$s_8 = -1$
$F_{-1}[7] =$	01.10000011	$F_1[7] = 10.01111011$	
w[8] =	10.01111111		
2w[8] =	100.11111110	$\widehat{y} = -4$	$s_9 = -1$
$F_{-1}[8] =$	01.1000001	$F_1[8] = 10.01111101$	
w[9] =	10.01111111		

We perform 9 iterations to compute the additional bit required for rounding. Since w[9] < 0 the correction step has to be performed. Thus $s_9 = -2$. The result is

$$s = 0.111\bar{1}\bar{1}\bar{1}\bar{1}\bar{1}\bar{1}\bar{1}\bar{2} = (0.11000000)_2$$

b) Radix-2, $s_j \in \{-1, 0, 1\}$, carry-save residual

2WS[0] =	0001.00100000	$\widehat{y} = 1$	$s_1 = 1$
2WC[0] =	0000.00000000		
	111.10000000	$F_{-1}[0] = 111.10000000$	
WS[1] =	110.10100000		
WC[1] =	010.00000000		
2WS[1] =	1101.01000000	$\widehat{y} = 1$	$s_2 = 1$
2WC[1] =	0100.00000000		
$F_1[1] =$	110.11000000	$F_{-1}[1] = 000.11000000$	
WS[2] =	111.10000000		
WC[2] =	000.10000000		
$2WS\left[2\right] =$	1111.00000000	$\widehat{y} = 0$	$s_3 = 1$
2WC[2] =	0001.00000000		
	110.01100000	$F_{-1}[2] = 001.01100000$	
WS[3] =	000.01100000		
WC[3] =	110.00000000		
$2WS\left[3\right] =$	0000.11000000	$\widehat{y} = -4$	$s_4 = -1$
2WC[3] =	1100.00000000		
$F_{-1}[3] =$	001.10110000	$F_1[3] = 110.00110000$	
WS[4] =	101.01110000		
WC[4] =	001.00000000		
,, 0 [1] -	001.00000000		

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2WS[4] =	1010.11100000	$\widehat{y} = -4$	$s_5 = -1$
$2WC\left[4\right] =$	0010.00000000		
$F_{-1}[4] =$	0010.00000000 001.10011000 001.01111000	$F_1[4] = 110.01011000$	
WS[5] =	001.01111000		
WC[5] =	101.00000000		
2WS[5] =	0010.11110000	$\widehat{y} = -4$	$s_6 = -1$
2WC[5] =	1010.00000000		
$F_{-1}[5] =$	001.10001100	$F_1[5] = 001.10001100$	
WS[6] =	001.10001100 001.01111100		
WC[6] =	101.00000000		
2WS[6] =	0010.11111000	$\widehat{y} = -4$	$s_7 = -1$
2WC[6] =	1010.00000000		
$F_{-1}[6] =$	001.10000110 001.01111110	$F_1[6] = 110.01110110$	
WS[7] =	001.01111110		
WC[7] =	101.00000000		
$2WS\left[7\right] =$	0010.11111100	$\widehat{y} = -4$	$s_8 = -1$
$2WC\left[7\right] =$	1010.00000000		
$F_{-1}[7] =$	001.10000011 001.01111111	$F_1[7] = 110.01111011$	
	001.01111111		
WC[8] =	101.00000000		
$2WS\left[8\right] =$	0010.11111110	$\widehat{y} = -4$	$s_9 = -1$
$2WC\left[8\right] =$	1010.00000000		
$F_{-1}[8] =$	001.10000001 001.01111111	$F_1[8] = 110.01111101$	
WS[9] =	001.01111111		
WC[9] =	101.00000000		

We perform 9 iterations to compute the additional bit required for rounding. Since w[9] < 0 the correction step has to be performed. Thus $s_9 = -2$. The result is

 $s = 0.111\overline{1}\overline{1}\overline{1}\overline{1}\overline{1}\overline{1}\overline{1}\overline{2} = (0.11000000)_2$

c) Radix-4, $s_j \in \{-2, -1, 0, 1, 2\}$, carry-save residual Since $\rho = \frac{a}{r-1} = \frac{2}{3} < 1$, s_0 should be 1. Therefore $w[0] = 1 - s_0 = 111.10010000$.

$4WS [0] =$ $4WC [0] =$ $F_1 [0] =$	1110.01000000 0000.00000000 001.11000000	$\hat{S} = 1.0000$ $\hat{y} = 1110.010$	S[0] = 1 $s_1 = -1$ S[1] = 0.11
WS [1] = WC [1] = $4WS [1] =$	11.10000000 00.10000000 1110.00000000	$\hat{S} = 0.1100$	$s_2 = 0$
$\frac{4WC[1] =}{WS[2] =}$ $WC[2] =$	0010.00000000 00.00000000 00.00000000	$\hat{y} = 0000.000$	S[2] = 0.1100
4WS [2] = 4WC [2] =	0000.00000000 0000.00000000	$\widehat{S} = 0.1100$ $\widehat{y} = 0000.000$	$s_3 = 0$ S[3] = 0.110000

Since w = 0, the rest of the digits of S are 0. We perform 4 iterations to take into account the generation of the additional bit required for rounding. The radix-4 digits of the result are $s_0 = 1$, $s_1 = -1$, $s_2 = 0$, $s_3 = 0$, $s_4 = 0$ and $s_5 = 0$. The result is

 $s = (0.11000000)_2$

Exercise 6.3

a) Use S[j] in its original signed digit form

In this case it is not necessary the on-the-fly conversion of S[j] for implementing the recurrence. Neverthless the register K[j] is still necessary. F[j] is computed as

$$-S_{j+1}\left(2S\left[j\right] + S_{j+1}r^{-(j+1)}\right)$$

which requires a single concatenation of S_{j+1} , and a digit multiplication by S_{j+1} . Since F[j] is represented in signed-digit form, the adder of the recurrence is more complex, that is, both operands are redundant.

b) Convert S[j] to two's complement representation

The conversion is on-the-fly, and since this conversion is already necessary, it does not introduce additional complexity. The adder is simpler that in a) since one operand is in nonredundant form. More specifically the term $-S_{j+1} \left(2S[j] + S_{j+1}r^{-(j+1)}\right)$ is generated in nonredundant form as follows:

 $-S_{j+1} \geq 0$

Concatenate S_{j+1} to 2S[j] in position j+1. Set the most significant digit to one to have a negative operand (the weight of the most significant digit is negative). Then perform digit multiplication.

 $-S_{j+1} < 0$

In this case

$$\left(2S[j] + S_{j+1}r^{-(j+1)}\right) = 2\left(S[j] - r^{-j}\right) + \left(2r - S_{j+1}\right)r^{-(j+1)}$$

The term $S[j] - r^{-j}$ is available from the on-the-fly conversion module. The term $2r - S_{j+1}$ is precomputed for every digit and is concatenated to $2(S[j] - r^{-j})$ in postion j + 1. Finally, the digit multiplication is performed.

Exercise 6.5

a) Network for digit selection

Figure E6.5a shows the network for the selection of s_{j+1} and s_{j+2} in a radix-2 square root implementation using two radix-2 overlapped stages.

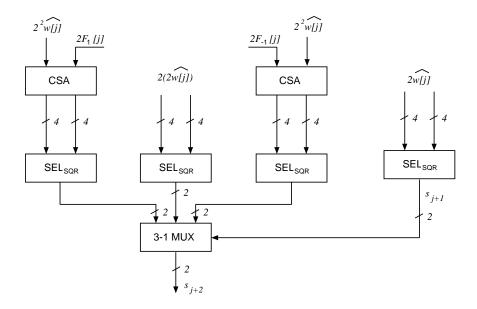


Figure E6.5a: Network for digit selection.

b) Network to produce the next residual

In Figure E6.5b the network producing the next residual is illustrated.

- c) Delay analysis
 - Conventional implementation

Computing the delay in the critical path we have

 $t_{cycle} = t_{SELSQRT}(4) + t_{buff}(1) + t_{mux}(1) + t_{HA}(1) + t_{reg}(2) = 9t_g$

The latency of the conventional implementation (8 fractional bits) can be computed as $8 \times t_{cycle} = 8 \times 9t_g = 72t_g$.

- Overlapped implementation

Computing the delay in the critical path we have that the delay to produce W[j+1] (that is, the delay from W[j] to W[j+1]) is

 $t_{SELSQRT}(4) + t_{buff}(1) + t_{mux}(1) + t_{HA}(1) = 7t_g$

Moreover, the delay to produce s_{j+2} (delay of CSA + delay of selection network + delay of 3-1 multiplexer) is

$$t_{CSA}(2) + t_{SELSQRT}(4) + t_{mux}(1) + t_{buff}(1) = 8t_g$$

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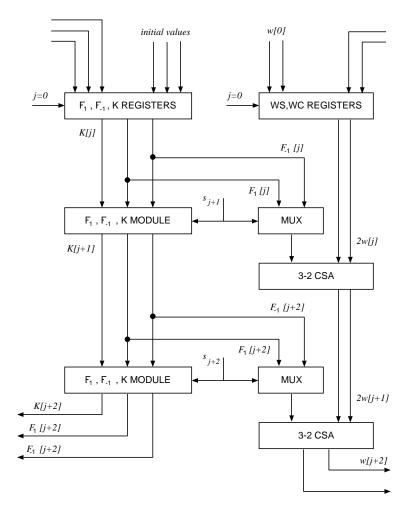


Figure E6.5b: Network to produce the next residual.

Finally, the delay to produce W[j+2] (delay to produce s_{j+2} + delay of buffer + delay of mux + delay of HA) can be computed as

$$8t_g + 1t_g + 1t_g + 1t_g = 12t_g$$

Adding the register delay we get $t_{cycle} = 11t_g + 2t_g = 13t_g$. Computing the latency of the overlapped implementation (8 fractional bits) we get $4 \times t_{cycle} = 4 \times 13t_g = 52t_g$

Exercise 6.8

We compute the radix-4 square root of $x = (53)_{10} = (00110101)_2$. Since n = 8, we perform a right-shift of m = 2 bits and produce $x^* = .11010100$.

The number of bits of the integer result is $\frac{8-2}{2} = 3$. Consequently, two radix-4 iterations are necessary. We have S[0] = 1 and $w[0] = x^* - 1 = 11.11010100$.

Note that no alignment to digit boundary is needed, since the square root algorithm does not require to compute a remainder.

The iterations are as follows:

4WS[0] =1111.01010000 4WC[0] =0000.00000000 $\hat{y} = 1111.0101$ $s_1 = -1$ S[1] = 0.11 $F_{-1}[0] =$ 001.11000000 WS[1] =10.10010000 WC[1] =10.10000000 4WS[1] =1010.01000000 $\hat{y} = 0100.0100$ $s_2 = 2$ S[2] = 0.11104WC[1] =1010.00000000

We do not need to compute w[2]. Therefore the result is

$$s = 2^3 (0.111) = 111 = (7)_{10}$$

Exercise 6.13

- k > 0

We develop a radix-4 selection function for J = 3, t = 3 and $\delta = 4$.

$$\min\left(U_{k-1}\left(I_{i}\right)\right) = 2 \times \left(\frac{1}{2} + i \times 2^{-4}\right) \times \left(k - \frac{1}{3}\right)$$
$$\max\left(L_{k}\left(I_{i}\right)\right) = 2 \times \left(\frac{1}{2} + (i+1) \times 2^{-4}\right) \times \left(k - \frac{2}{3}\right)$$

 $-k \leq 0$

$$\min(U_{k-1}(I_i)) = 2 \times \left(\frac{1}{2} + (i+1) \times 2^{-4}\right) \times \left(k - \frac{1}{3}\right)$$

$$\max\left(L_{k}\left(I_{i}\right)\right) = 2 \times \left(\frac{1}{2} + i \times 2^{-4}\right) \times \left(k - \frac{2}{3}\right) + \left(k - \frac{2}{3}\right)^{2} \times 4^{-4}$$

$$\widehat{L}_{k} = \max\left(\left\lceil L_{k}\left(I_{i}\right)\right\rceil_{3}\right) \le m_{k}\left(i\right) \le \min\left(\left\lfloor U_{k-1}\left(I_{i}\right)\right) - 2^{-3}\right\rfloor_{3} = \widehat{U}_{k-1}$$

To improve the presentation of results, we use a bound for max $(L_k(I_i))$. More specifically, we want an upper bound of the term $(k - \frac{2}{3})^2 \times 4^{-4}$. For k = 0 we have $\frac{4}{9} \times 4^{-4} = \frac{1}{576} < \frac{1}{512}$. For k = 1 we have $(-\frac{5}{3})^2 \times 4^{-4} = \frac{25}{2304} < \frac{1}{64}$. The selection constants are presented in Table E6.13. Note that we give only held of the table (for $\widehat{G}(k) = 0.210$.

The selection constants are presented in Table E6.13. Note that we give only half of the table (for $\hat{S}[j] = 8, 9, 10, 11$) since there is an interval $\hat{U}_{-2} - \hat{L}_{-1}$ that is negative. Consequently, there is no selection function for t = 3 and $\delta = 4$.

$\widehat{S}\left[j ight]$	8	9	10	11
$\widehat{L}_2, \widehat{U}_1$	12, 12	14, 14	$15, \ 15$	16, 17
m_2	12	14	15	16
$\widehat{L}_1, \widehat{U}_0$	3, 4	4, 5	4, 5	4, 6
m_1	4	4	4	4
$\widehat{L}_0, \widehat{U}_{-1}$	-5, -4	-5, -5	-6, -5	-7, -5
m_0	-4	-5	-6	-6
$\widehat{L}_{-1}, \widehat{U}_{-2}$	-13, -13	-14, -15	-16, -16	-18, -17
m_{-1}	-13	X	-16	-18

Table E6.13: Selection interval and m_k constants. $\widehat{S}[j]$: real value= shown value/16. \widehat{L}_k , \widehat{U}_{k-1} and m_k : real value = shown value/8.